# Linear System of Equations 

## An Introduction - A World in Notation

Systems of linear equations are common in science and mathematics. A system of linear equations (or linear system) is a collection of one or more linear equations involving the same set of variables.

## A Linear equation

- There are problems where their solution lead us in an equation with two unknowns $x, y$, of the following form

$$
a x+b y=c
$$

- Solution of the equation $a x+b y=c$ is called any pair of numbers $(x, y)$ that verifies it.


## $1^{\text {st }}$ Example

- The equation $2 x+y=6$
- We note that for $x=1$ and $y=4$ the equation is verified, but for $\mathrm{x}=2$ and $\mathrm{y}=1$ is not verified.
- The pair of numbers $(1,4)$ that verifies the equation $2 x+y=6$ we say that is one of its solution.
- The equation $2 x+y=6$ does not have only the solution (1, 4), but it has infinite solutions.



## $2^{\text {nd }}$ Example

- Suppose we have two linear equations with
two unknowns

$$
\begin{gathered}
x+y=5 \\
2 x+y=8
\end{gathered}
$$

- Solution of a linear system of two equations
with two unknowns $x$ and $y$ is called each pair $(x, y)$ that verifies its equations.


## Graphic solution of a linear system with two unknowns

- For the graphical solution of a $2 \times 2$ linear system we work as follows

$$
\begin{gathered}
x+y=5 \\
2 x+y=8
\end{gathered}
$$

- We draw in Geogebra the lines in the same axis system

$$
\begin{aligned}
& \operatorname{Eq}(1): \quad x+y=5 \\
& E q(2): 2 x+y=8
\end{aligned}
$$



- We observe that the lines intersect at the point $(\mathbf{3}, \mathbf{2})$. This point belongs to both lines and the coordinates of $x=3$ and $y=2$ verify both equations of the system. So the pair $(3,2)$ is a solution of the system.
- These lines have no other common point, so the system has no other solution. This means that the pair $(3,2)$ is the only solution in the system.


## $3^{\text {rd }}$ Example

- Suppose now that we have the following system:

$$
\begin{gathered}
2 x-3 y=6 \\
4 x-6 y=-24
\end{gathered}
$$

- To solve the system we draw the lines. We notice in the next figure that the lines are parallel.
- This means that they have no common point, so the system has no solution. In this case we say that the system is impossible.


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## $4^{\text {th }}$ Example

- Suppose now that we have the following system:

$$
\begin{gathered}
3 x-y=6 \\
6 x-2 y=12
\end{gathered}
$$

- To solve the system we draw the lines. We notice in the figure on the right, that the lines are the same.
- So they have all their points in common and therefore the system has infinite solutions.



## Algebraic solution of a linear system

## Elimination of variables ( $\mathbf{2 x 2}$ system)

The simplest method for solving a system of linear equations is to repeatedly eliminate variables. This method can be described as follows:

- In the first equation, solve for one of the variables in terms of the others.
- Substitute this expression into the other equation. Thus, a single linear equation.
- Solve this equation, and then back-substitute into the previous equation.
- The solution is found.


## 5th Example

- Solve the following system of linear equations

$$
\begin{gathered}
x+y=20 \\
x+3 y=44
\end{gathered}
$$

- Solve the equation $x+y=20$ with respect to $x$,
- So we have $x=20-y$
- We substitute $x$ with $20-y$ into the other equation $x+3 y=44$
$(20-y)+3 y=44$
$20+2 y=44$
$2 y=44-20$
$2 y=24$ thus $y=12$
For $y=12$ and from $x=20-y$ we obtain
$x=20-12$
$x=8$


## Verification

- So the solution is $x=8$ and $y=12$
- For verification we substitute the values of $x$ and $y$ into original system of equations

$$
\begin{gathered}
x+y=20 \\
x+3 y=44
\end{gathered}
$$

We get

$$
\begin{gathered}
8+12=20 \text { and } \\
8+3 * 12=8+36=44
\end{gathered}
$$

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## Algebraic solution of a linear system

## Elimination of variables (Generalization in nxn system)

The simplest method for solving a system of linear equations is to repeatedly eliminate variables. This method can be described as follows:

- In the first equation, solve for one of the variables in terms of the others.
- Substitute this expression into the remaining equations. This yields a system of equations with one fewer equation and one fewer unknown.
- Repeat until the system is reduced to a single linear equation.
- Solve this equation, and then back-substitute until the entire solution is found.


## $6^{\text {th }}$ Example

The system

$$
\begin{aligned}
3 x_{1}+2 x_{2}+x_{3} & =1 \\
x_{2}-x_{3} & =2 \\
2 x_{3} & =4
\end{aligned}
$$

is in strict triangular form, since in the second equation the coefficients are $0,1,-1$, respectively, and in the third equation the coefficients are $0,0,2$, respectively. Because of the strict triangular form, the system is easy to solve. It follows from the third equation that $x_{3}=2$. Using this value in the second equation, we obtain

$$
x_{2}-2=2 \quad \text { or } \quad x_{2}=4
$$

Using $x_{2}=4, x_{3}=2$ in the first equation, we end up with

$$
\begin{aligned}
3 x_{1}+2 \cdot 4+2 & =1 \\
x_{1} & =-3
\end{aligned}
$$

Thus, the solution of the system is $(-3,4,2)$.

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## $7^{\text {th }}$ Example from Electrical Circuits

Solve for the current flowing through the $8 \Omega$ resistor

Therefore, we will solve for $i_{8 \Omega}$ using nodal analysis.
4. Attempt a problem solution. We first write down all of the equations we will need in order to find $i_{8 \Omega}$.

$$
\begin{aligned}
& i_{8 \Omega}=i_{2}, \quad i_{2}=\frac{v_{1}}{8}, \quad i_{8 \Omega}=\frac{v_{1}}{8} \\
& \frac{v_{1}-5}{2}+\frac{v_{1}-0}{8}+\frac{v_{1}+3}{4}=0
\end{aligned}
$$

Now we can solve for $v_{1}$.

$$
\begin{gathered}
8\left[\frac{v_{1}-5}{2}+\frac{v_{1}-0}{8}+\frac{v_{1}+3}{4}\right]=0 \\
\text { leads to }\left(4 v_{1}-20\right)+\left(v_{1}\right)+\left(2 v_{1}+6\right)=0 \\
7 v_{1}=+14, \quad v_{1}=+2 \mathrm{~V}, \quad i_{8 \Omega}=\frac{v_{1}}{8}=\frac{2}{8}=\mathbf{0 . 2 5} \mathbf{~ A}
\end{gathered}
$$


5. Evaluate the solution and check for accuracy. We can now use Kirchhoff's voltage law (KVL) to check the results.

$$
\begin{gathered}
i_{1}=\frac{v_{1}-5}{2}=\frac{2-5}{2}=-\frac{3}{2}=-1.5 \mathrm{~A} \\
i_{2}=i_{8 \Omega}=0.25 \mathrm{~A} \\
i_{3}=\frac{v_{1}+3}{4}=\frac{2+3}{4}=\frac{5}{4}=1.25 \mathrm{~A} \\
i_{1}+i_{2}+i_{3}=-\mathbf{1 . 5}+\mathbf{0 . 2 5}+\mathbf{1 . 2 5}=\mathbf{0}
\end{gathered} \text { (Checks.) }
$$

## Co-funded by the Erasmus+ Programme of the European Union <br> $8^{\text {th }}$ Example from Electrical Circuits

Assume an electric network consisting of two voltage sources and three resistors.
According to the first law:

$$
i_{1}-i_{2}-i_{3}=0
$$

Applying the second law to the closed circuit $s_{1}$, and substituting for voltage using Ohm's law gives:

$$
-R_{2} i_{2}+\mathcal{E}_{1}-R_{1} i_{1}=0
$$

The second law, again combined with Ohm's law, applied to the closed circuit $s_{2}$ gives:

$$
-R_{3} i_{3}-\mathcal{E}_{2}-\mathcal{E}_{1}+R_{2} i_{2}=0
$$

This yields a system of linear equations in $i_{1}, i_{2}, i_{3}$ :

$$
\begin{cases}i_{1}-i_{2}-i_{3} & =0 \\ -R_{2} i_{2}+\mathcal{E}_{1}-R_{1} i_{1} & =0 \\ -R_{3} i_{3}-\mathcal{E}_{2}-\mathcal{E}_{1}+R_{2} i_{2} & =0\end{cases}
$$


which is equivalent to

$$
\begin{cases}i_{1}+\left(-i_{2}\right)+\left(-i_{3}\right) & =0 \\ R_{1} i_{1}+R_{2} i_{2}+0 i_{3} & =\mathcal{E}_{1} \\ 0 i_{1}+R_{2} i_{2}-R_{3} i_{3} & =\mathcal{E}_{1}+\mathcal{E}_{2}\end{cases}
$$

Assuming

$$
\begin{aligned}
& R_{1}=100 \Omega, R_{2}=200 \Omega, R_{3}=300 \Omega \\
& \mathcal{E}_{1}=3 \mathrm{~V}, \mathcal{E}_{2}=4 \mathrm{~V}
\end{aligned}
$$

- Thus the system of linear equation becomes

$$
\begin{gathered}
i_{1}-\quad i_{2}-\quad i_{3}=0 \\
100 \cdot i_{1}+200 \cdot i_{2}+3 \cdot i_{3}=3 \\
0 \cdot i_{1}+200 \cdot i_{2}-300 \cdot i_{3}=7 \\
\Delta=\left|\begin{array}{ccc}
1 & -1 & -1 \\
100 & 200 & 0 \\
0 & 200 & -300
\end{array}\right|=-110000
\end{gathered}
$$

$$
\begin{aligned}
& \Delta_{1}=\left|\begin{array}{ccc}
0 & -1 & -1 \\
3 & 200 & 0 \\
7 & 200 & -300
\end{array}\right|=-100 ; \\
& \Delta_{2}=\left|\begin{array}{ccc}
1 & 0 & -1 \\
100 & 3 & 0 \\
0 & 7 & -300
\end{array}\right|=-1600
\end{aligned}
$$

$$
\begin{aligned}
& \Delta_{3}=\left|\begin{array}{ccc}
1 & -1 & 0 \\
100 & 200 & 3 \\
0 & 200 & 7
\end{array}\right|=1500 \\
& x_{1}=\Delta_{1} / \Delta=\frac{-100}{-110000}=\frac{1}{1100} \\
& x_{2}=\Delta_{2} / \Delta=\frac{-1600}{-110000}=\frac{4}{275} \\
& x_{3}=\Delta_{3} / \Delta=\frac{1500}{-110000}=\frac{-3}{220}
\end{aligned}
$$

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the solution is

$$
\left\{\begin{array}{l}
i_{1}=\frac{1}{1100} \mathrm{~A} \\
i_{2}=\frac{4}{275} \mathrm{~A} \\
i_{3}=-\frac{3}{220} \mathrm{~A}
\end{array}\right.
$$

The current $i_{3}$ has a negative sign which means the assumed direction of $i_{3}$ was incorrect and $i_{3}$ is actually flowing in the direction opposite to the red arrow labeled $i_{3}$. The current in $R_{3}$ flows from left to right.

## 9th Example

Solve the system

$$
\begin{aligned}
x_{1}+2 x_{2}+x_{3}= & 3 \\
3 x_{1}-x_{2}-3 x_{3}= & -1 \\
2 x_{1}+3 x_{2}+x_{3}= & 4
\end{aligned}
$$

## Solution

Subtracting 3 times the first row from the second row yields

$$
-7 x_{2}-6 x_{3}=-10
$$

Subtracting 2 times the first row from the third row yields

$$
-x_{2}-x_{3}=-2
$$

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If the second and third equations of our system, respectively, are replaced by these new equations, we obtain the equivalent system

$$
\begin{aligned}
x_{1}+2 x_{2}+x_{3} & =3 \\
-7 x_{2}-6 x_{3} & =-10 \\
-x_{2}-x_{3} & =-2
\end{aligned}
$$

If the third equation of this system is replaced by the sum of the third equation and $-\frac{1}{7}$ times the second equation, we end up with the following strictly triangular system:

$$
\begin{aligned}
x_{1}+2 x_{2}+x_{3} & =3 \\
-7 x_{2}-6 x_{3} & =-10 \\
-\frac{1}{7} x_{3} & =-\frac{4}{7}
\end{aligned}
$$

Using back substitution, we get

$$
x_{3}=4, \quad x_{2}=-2, \quad x_{1}=3
$$

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## Matrix

- A matrix is a rectangular array of numbers or other mathematical objects for which operations such as addition and multiplication are defined.
- Most commonly, a matrix over a field $F$ is a rectangular array of scalars each of which is a member of $F$.
- Here we focuses on real and complex matrices, that is, matrices whose elements are real numbers or complex numbers, respectively. For instance, this is a real matrix:

$$
\left(\begin{array}{rrr}
1 & 2 & 1 \\
3 & -1 & -3 \\
2 & 3 & 1
\end{array}\right)
$$

- The numbers, symbols, or expressions in the matrix are called its entries or its elements. The horizontal and vertical lines of entries in a matrix are called rows and columns, respectively.

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## Types of Matrices

| Row <br> vector | $1 \times n$ | $\left[\begin{array}{lll}3 & 7 & 2\end{array}\right]$ | A matrix with one row, sometimes used to represent a vector |
| :---: | :---: | :---: | :---: |
| Column <br> vector | $n \times 1$ | $\left[\begin{array}{l}4 \\ 1 \\ 8\end{array}\right]$ | A matrix with one column, sometimes used to represent a vector |
| Square <br> matrix | $n \times n$ | $\left[\begin{array}{ccc}9 & 13 & 5 \\ 1 & 11 & 7 \\ 2 & 6 & 3\end{array}\right]$ |  |

1
2
3
$\vdots$
$m$$\left[\begin{array}{cccc}1 & 2 & \ldots & n \\ a_{11} & a_{12} & \ldots & a_{1 n} \\ a_{21} & a_{22} & \ldots & a_{2 n} \\ a_{31} & a_{32} & \ldots & a_{3 n} \\ \vdots & \vdots & \vdots & \vdots \\ a_{m 1} & a_{m 2} & \ldots & a_{m n}\end{array}\right]$

An $m \times n$ matrix: the $m$ rows are horizontal and the $n$ columns are vertical. Each element of a matrix is often denoted by a variable with two subscripts. For example, $a_{2,1}$ represents the element at the second row and first column of the matrix.

## Addition and Scalar multiplication

The sum $\mathbf{A}+\mathbf{B}$ of two m-by-n matrices $\mathbf{A}$ and $\mathbf{B}$ is calculated

$$
\begin{array}{l|l}
\text { Addition } & \text { entrywise: } \\
\quad(\mathbf{A}+\mathbf{B})_{i, j}=\mathbf{A}_{i, j}+\mathbf{B}_{i, j}, \text { where } 1 \leq i \leq m \text { and } 1 \leq j \leq n .
\end{array}
$$

$$
\left[\begin{array}{lll}
1 & 3 & 1 \\
1 & 0 & 0
\end{array}\right]+\left[\begin{array}{lll}
0 & 0 & 5 \\
7 & 5 & 0
\end{array}\right]=\left[\begin{array}{lll}
1+0 & 3+0 & 1+5 \\
1+7 & 0+5 & 0+0
\end{array}\right]=\left[\begin{array}{lll}
1 & 3 & 6 \\
8 & 5 & 0
\end{array}\right]
$$

The product $c \mathbf{A}$ of a number $c$ (also called a scalar in the parlance of abstract algebra) and a matrix $\mathbf{A}$ is computed by
multiplying every entry of $\mathbf{A}$ by $c$ :

$$
(c \mathbf{A})_{i, j}=c \cdot \mathbf{A}_{i, j} .
$$

$$
2 \cdot\left[\begin{array}{ccc}
1 & 8 & -3 \\
4 & -2 & 5
\end{array}\right]=\left[\begin{array}{ccc}
2 \cdot 1 & 2 \cdot 8 & 2 \cdot-3 \\
2 \cdot 4 & 2 \cdot-2 & 2 \cdot 5
\end{array}\right]=\left[\begin{array}{ccc}
2 & 16 & -6 \\
8 & -4 & 10
\end{array}\right]
$$

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## Matrix Transposition

Transposition $\left\lvert\,$\begin{tabular}{l}
The transpose of an $m$-by- $n$ matrix $\mathbf{A}$ is the $n$-by- $m$ matrix $\mathbf{A}^{\top}$ <br>

| (also denoted $\mathbf{A}^{\text {tr }}$ |
| :--- |
| and vice versa: |
|  |
| $\quad\left(\mathbf{A}^{\top}\right)_{i, j}=\mathbf{A}_{j, i}$. |

\end{tabular}\(\quad\left[\begin{array}{ccc}1 \& 2 \& 3 <br>

0 \& -6 \& 7\end{array}\right]^{\mathrm{T}}=\left[$$
\begin{array}{cc}1 & 0 \\
2 & -6 \\
3 & 7\end{array}
$$\right]\right.\)

Familiar properties of numbers extend to these operations of matrices: for example, addition is commutative, that is, the matrix sum does not depend on the order of the summands: $A+B=B+A$.
The transpose is compatible with addition and scalar multiplication, as expressed by $(c A)^{\top}=c\left(A^{\top}\right)$ and $(A+B)^{\top}=$ $A^{\top}+B^{\top}$. Finally, $\left(A^{\top}\right)^{\top}=A$.

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## Matrix multiplication

- Multiplication of two matrices is defined if and only if the number of columns of the left matrix is the same as the number of rows of the right matrix.
- If $\mathbf{A}$ is an $m$-by- $n$ matrix and $\mathbf{B}$ is an $n$-by- $p$ matrix, then their matrix product $\mathbf{A B}$ is the $m$-by- $p$ matrix whose entries are given by dot product of the corresponding row of $\mathbf{A}$ and the corresponding column of B:

$[\mathbf{A B}]_{i, j}=a_{i, 1} b_{1, j}+a_{i, 2} b_{2, j}+\cdots+a_{i, n} b_{n, j}=\sum_{r=1}^{n} a_{i, r} b_{r, j}$,

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$$
\left[\begin{array}{lll}
\underline{2} & \underline{3} & \underline{4} \\
1 & 0 & 0
\end{array}\right]\left[\begin{array}{ll}
0 & \frac{1000}{100} \\
1 & \underline{10}
\end{array}\right]=\left[\begin{array}{ll}
3 & \underline{2340} \\
0 & \underline{1000}
\end{array}\right]
$$

- Matrix multiplication satisfies the rules $(\mathbf{A B}) \mathbf{C}=\mathbf{A}(\mathbf{B C})$ (associativity), and $(\mathbf{A}+\mathbf{B}) \mathbf{C}=\mathbf{A C}+\mathbf{B C}$ as well as $\mathbf{C}(\mathbf{A}+\mathbf{B})$ $=\mathbf{C A}+\mathbf{C B}$ (left and right distributivity), whenever the size of the matrices is such that the various products are defined.
- The product $\mathbf{A B}$ may be defined without $\mathbf{B A}$ being defined, namely if $\mathbf{A}$ and $\mathbf{B}$ are $m-b y-n$ and $n-b y-k$ matrices, respectively, and $m \neq k$. Even if both products are defined, they need not be equal, that is, generally
- $A B \neq B A$
- that is, matrix multiplication is not commutative, in marked contrast to (rational, real, or complex) numbers whose product is independent of the order of the factors. An example of two matrices not commuting with each other is:
- An example of two matrices not commuting with each other is:

$$
\left[\begin{array}{ll}
1 & 2 \\
3 & 4
\end{array}\right]\left[\begin{array}{ll}
0 & 1 \\
0 & 0
\end{array}\right]=\left[\begin{array}{ll}
0 & 1 \\
0 & 3
\end{array}\right]
$$

whereas

$$
\left[\begin{array}{ll}
0 & 1 \\
0 & 0
\end{array}\right]\left[\begin{array}{ll}
1 & 2 \\
3 & 4
\end{array}\right]=\left[\begin{array}{ll}
3 & 4 \\
0 & 0
\end{array}\right]
$$

## Submatrix

- A submatrix of a matrix is obtained by deleting any collection of rows and/or columns.
- For example, from the following 3-by-4 matrix, we can construct a 2-by-3 submatrix by removing row 3 and column 2:

$$
\mathbf{A}=\left[\begin{array}{cccc}
1 & 2 & 3 & 4 \\
5 & 6 & 7 & 8 \\
9 & 10 & 11 & 12
\end{array}\right] \rightarrow\left[\begin{array}{ccc}
1 & 3 & 4 \\
5 & 7 & 8
\end{array}\right]
$$

## Square matrix

- A square matrix is a matrix with the same number of rows and columns. An n-by$n$ matrix is known as a square matrix of order $n$. Any two square matrices of the same order can be added and multiplied. The entries $a_{i i}$ form the main diagonal of a square matrix. They lie on the imaginary line that runs from the top left corner to the bottom right corner of the matrix.


## Diagonal and triangular matrix

- If all entries of $\mathbf{A}$ below the main diagonal are zero, $\mathbf{A}$ is called an upper triangular matrix.
- Similarly if all entries of $A$ above the main diagonal are zero, $\mathbf{A}$ is called a lower triangular matrix.
- If all entries outside the main diagonal are zero, $\mathbf{A}$ is called a diagonal matrix.

| Name | Example with $\boldsymbol{n}=3$ |
| :---: | :---: |
| Diagonal matrix | $\left[\begin{array}{ccc}a_{11} & 0 & 0 \\ 0 & a_{22} & 0 \\ 0 & 0 & a_{33}\end{array}\right]$ |
| Lower triangular matrix | $\left[\begin{array}{ccc}a_{11} & 0 & 0 \\ a_{21} & a_{22} & 0 \\ a_{31} & a_{32} & a_{33}\end{array}\right]$ |
| Upper triangular matrix | $\left[\begin{array}{ccc}a_{11} & a_{12} & a_{13} \\ 0 & a_{22} & a_{23} \\ 0 & 0 & a_{33}\end{array}\right]$ |

## Identity matrix

- The identity matrix $\mathrm{I}_{n}$ of size $n$ is the $n$-by- $n$ matrix in which all the elements on the main diagonal are equal to 1 and all other elements are equal to 0 , for example,

$$
\mathbf{I}_{1}=[1], \mathbf{I}_{2}=\left[\begin{array}{ll}
1 & 0 \\
0 & 1
\end{array}\right], \cdots, \mathbf{I}_{n}=\left[\begin{array}{cccc}
1 & 0 & \cdots & 0 \\
0 & 1 & \cdots & 0 \\
\vdots & \vdots & \ddots & \vdots \\
0 & 0 & \cdots & 1
\end{array}\right]
$$

## Invertible matrix and its inverse

- A square matrix $\mathbf{A}$ is called invertible or non-singular if there exists a matrix $\mathbf{B}$ such that
- $A B=B A=I_{n}$
- where $\mathrm{I}_{\mathrm{n}}$ is the $\mathrm{n} \times \mathrm{n}$ identity matrix with 1 s on the main diagonal and 0 s elsewhere. If $\mathbf{B}$ exists, it is unique and is called the inverse matrix of $\mathbf{A}$, denoted $\mathrm{A}^{\mathbf{- 1}}$.


## Determinant

- The determinant $\operatorname{det}(\mathbf{A})$ or $|\mathbf{A}|$ of a square matrix $\mathbf{A}$ is a number encoding certain properties of the matrix.
- A matrix is invertible if and only if its determinant is nonzero.


## Determinant

In the case of a $2 \times 2$ matrix the determinant may be defined as

$$
|A|=\left|\begin{array}{ll}
a & b \\
c & d
\end{array}\right|=a d-b c .
$$

Similarly, for a $3 \times 3$ matrix $A$, its determinant is

$$
\begin{aligned}
|A|=\left|\begin{array}{ccc}
a & b & c \\
d & e & f \\
g & h & i
\end{array}\right| & =a\left|\begin{array}{ccc}
\square & \square & \square \\
\square & e & f \\
\square & h & i
\end{array}\right|-b\left|\begin{array}{ccc}
\square & \square & \square \\
d & \square & f \\
g & \square & i
\end{array}\right|+c\left|\begin{array}{ccc}
\square & \square & \square \\
d & e & \square \\
g & h & \square
\end{array}\right| \\
& =a\left|\begin{array}{cc}
e & f \\
h & i
\end{array}\right|-b\left|\begin{array}{cc}
d & f \\
g & i
\end{array}\right|+c\left|\begin{array}{cc}
d & e \\
g & h
\end{array}\right| \\
& =a e i+b f g+c d h-c e g-b d i-a f h
\end{aligned}
$$

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## General form

A general system of $m$ linear equations with $n$ unknowns can be written as

$$
\begin{aligned}
a_{11} x_{1}+a_{12} x_{2}+\cdots+a_{1 n} x_{n} & =b_{1} \\
a_{21} x_{1}+a_{22} x_{2}+\cdots+a_{2 n} x_{n} & =b_{2} \\
& \vdots \\
a_{m 1} x_{1}+a_{m 2} x_{2}+\cdots+a_{m n} x_{n} & =b_{m}
\end{aligned}
$$

where $x_{1}, x_{2}, \ldots, x_{n}$ are the unknowns, $a_{11}, a_{12}, \ldots, a_{m n}$ are the coefficients of the system, and $b_{1}, b_{2}, \ldots, b_{m}$ are the constant terms.

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## Matrix equation

The vector equation is equivalent to a matrix equation of the form

$$
A \mathbf{x}=\mathbf{b}
$$

where $A$ is an $m \times n$ matrix, $\mathbf{x}$ is a column vector with $n$ entries, and $\mathbf{b}$ is a column vector with $m$ entries.

$$
A=\left[\begin{array}{cccc}
a_{11} & a_{12} & \cdots & a_{1 n} \\
a_{21} & a_{22} & \cdots & a_{2 n} \\
\vdots & \vdots & \ddots & \vdots \\
a_{m 1} & a_{m 2} & \cdots & a_{m n}
\end{array}\right], \quad \mathbf{x}=\left[\begin{array}{c}
x_{1} \\
x_{2} \\
\vdots \\
x_{n}
\end{array}\right], \quad \mathbf{b}=\left[\begin{array}{c}
b_{1} \\
b_{2} \\
\vdots \\
b_{m}
\end{array}\right]
$$

## Cramer's rule

$$
\begin{array}{r}
x+3 y-2 z=5 \\
3 x+5 y+6 z=7 \\
2 x+4 y+3 z=8
\end{array}
$$

is given by

$$
x=\frac{\left|\begin{array}{ccc}
5 & 3 & -2 \\
7 & 5 & 6 \\
8 & 4 & 3
\end{array}\right|}{\left|\begin{array}{ccc}
1 & 3 & -2 \\
3 & 5 & 6 \\
2 & 4 & 3
\end{array}\right|}, \quad y=\frac{\left|\begin{array}{ccc}
1 & 5 & -2 \\
3 & 7 & 6 \\
2 & 8 & 3
\end{array}\right|}{\left|\begin{array}{ccc}
1 & 3 & -2 \\
3 & 5 & 6 \\
2 & 4 & 3
\end{array}\right|}, \quad z=\frac{\left|\begin{array}{ccc}
1 & 3 & 5 \\
3 & 5 & 7 \\
2 & 4 & 8
\end{array}\right|}{\left|\begin{array}{ccc}
1 & 3 & -2 \\
3 & 5 & 6 \\
2 & 4 & 3
\end{array}\right|} .
$$

## Multipath propagation leads to ISI



## Zero-forcing equalizer



- We will assume that the filter has $(2 N+1)$ taps with gains $C_{-N}, C_{-N+1}, \ldots, C_{0}, C_{1}, \ldots$ $C_{N}$. The input of the equalizer is $p_{r}(t)$ that is known, and the output is $p_{e q}(t)$. We can write the output $p_{e q}(t)$ as a function of $p_{r}(t)$ and the gains of the taps as

$$
\begin{gathered}
p_{e q}(k)=\sum_{n=-N}^{N} C_{n} p_{r}(k-n) \\
p_{e q}(k)=\left\{\begin{array}{lr}
1 & \text { for } k=0 \\
0 & \text { for } k= \pm 1, \pm 2, \ldots, \pm N
\end{array}\right.
\end{gathered}
$$

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- By combining the equations above, we have
$\left[\begin{array}{c}0 \\ 0 \\ \vdots \\ 0 \\ 1 \\ 0 \\ \vdots \\ 0\end{array}\right]=\left[\begin{array}{cccc}p_{r}(0) & p_{r}(-1) & \cdots & p_{r}(-2 N) \\ p_{r}(1) & p_{r}(0) & \cdots & p_{r}(-2 N+1) \\ \vdots & & & \vdots \\ \vdots & & & \vdots \\ \vdots & & & \vdots \\ \vdots & & & \vdots \\ \vdots & & & \vdots \\ p_{r}(2 N) & \cdots & \cdots & p_{r}(0)\end{array}\right]\left[\begin{array}{c}C_{-N} \\ C_{-N+1} \\ \vdots \\ \vdots \\ C_{0} \\ \\ C_{N-1} \\ C_{N}\end{array}\right]$

This equation represents a system of $(2 N+1)$ equations and can be solved with respect to $C_{n}$. The equalizer describing the system is called a zero forcing equalizer because $p_{\text {eq }}(k)$ has $N$ zero values on each side.

## Example: How the 3 tap equalizer works

- Design a three-tap equalizer to reduce the ISI due to the non equalized pulse $\mathrm{p}_{\mathrm{r}}(\mathrm{t})$ of the figure below



## Solution

- At $t=0$, the current pulse is sampled, after $T_{s}$ sec the next sampling pulse will be taken, etc. Thus, we want the current pulse to be zero in the integer multiples of $\mathrm{T}_{\mathrm{s}}$ so that there are no residual current pulses at the next points sampling. That's what we want to apply to every pulse.
- Observing the non-equalized pulse and the equalized below, we see that this equalizer forces the pulse to zero at the sampling points of the next $T_{s}$ and the previous pulse $-T_{s}$, because it is three tap equalizer. If he was five tap equalizer, it would force the pulse to become zero at the $T_{s}, 2 T_{s}$ and $-T_{s^{\prime}}-2 T_{s}$ points.

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- With a three tap equalizer here, we can cause a "zeroing" in the equalized pulse both before and after the point $t=0$. The gains of the taps for this equalizer are given by the solution of the following $3 \times 3$ system of equations which is given below with the form of a matrix equation

$$
\left[\begin{array}{l}
0 \\
1 \\
0
\end{array}\right]=\left[\begin{array}{ccc}
1.0 & 0.1 & 0 \\
-0.2 & 1.0 & 0.1 \\
0.1 & -0.2 & 1
\end{array}\right]\left[\begin{array}{l}
C_{-1} \\
C_{0} \\
C_{1}
\end{array}\right]
$$

$$
\begin{aligned}
& \Delta=\left|\begin{array}{ccc}
1 & \frac{1}{10} & 0 \\
\frac{-1}{5} & 1 & \frac{1}{10} \\
\frac{1}{10} & \frac{-1}{5} & 1
\end{array}\right|=\frac{1041}{1000} \\
& \Delta_{2}=\left|\begin{array}{ccc}
1 & 0 & 0 \\
\frac{-1}{5} & 1 & \frac{1}{10} \\
\frac{1}{10} & 0 & 1
\end{array}\right|=1 \\
& x_{1}=\Delta_{1} / \Delta=\frac{\frac{-1}{\frac{10}{1041}}}{\frac{-100}{1000}}=\frac{-1041}{104} \\
& x_{2}=\Delta_{2} / \Delta=\frac{1}{\frac{1041}{1000}}=\frac{1000}{1041} \\
& \Delta_{1}=\left|\begin{array}{ccc}
0 & \frac{1}{10} & 0 \\
1 & 1 & \frac{1}{10} \\
0 & \frac{-1}{5} & 1
\end{array}\right|=\frac{-1}{10} ; \\
& \Delta_{3}=\left|\begin{array}{ccc}
1 & \frac{1}{10} & 0 \\
\frac{-1}{5} & 1 & 1 \\
\frac{1}{10} & \frac{-1}{5} & 0
\end{array}\right|=\frac{21}{100} ; \\
& x_{3}=\Delta_{3} / \Delta=\frac{\frac{21}{100}}{\frac{1041}{1000}}=\frac{70}{347}
\end{aligned}
$$

$$
\left[\begin{array}{l}
C_{-1} \\
C_{0} \\
C_{1}
\end{array}\right]=\left[\begin{array}{c}
-0.09606 \\
0.9606 \\
0.2017
\end{array}\right]
$$

- So, for these specific values of $\mathrm{C}_{-1}, \mathrm{C}_{0}$ and $\mathrm{C}_{1}$ the three tap equalizer; equalize the pulse in order to reduce ISI.

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## Thank you!

## Any Questions?

